

## SECTION I—CLAIMS

### **Amendment to the Claims:**

This listing of the claims will replace all prior versions and listings of claims in the application. Claim 1 is amended herein. Claims 1-29 remain canceled herein without prejudice. New claims 45-49 are presented herein.

### **Listing of Claims:**

1-29. (Canceled).

30. (Currently amended) A method comprising:

receiving content, at a diversity agent, the content for transmission from a wireless

communication system having  $M$  transmit antennae and  $N$  receive antennae and  $N_c$  subcarriers, where  $N_c \gg M, N$ , the received content for transmission from more than two of the  $M$  transmit antennae, wherein the received content is a vector of input symbols ( $\mathbf{s}$ ) of size  $N_c \times 1$ , and wherein the  $N_c$  subcarriers is the number of subcarriers of a multicarrier wireless communication channel of the wireless communication system; and generating a rate-one, space-frequency code matrix from the received content for transmission via the more than two of the  $M$  transmit antennae by dividing the vector of input symbols into a number  $G$  of groups to generate subgroups and multiplying at least a subset of the subgroups by a constellation rotation precoder to produce a number  $G$  of pre-coded vectors ( $\mathbf{v}_g$ ), wherein successive symbols from the same group transmitted from the same antenna are at a frequency distance that is multiples of  $MG$  subcarrier spacings.

31. (Previously Presented) A method according to claim 30, further comprising:

dividing each of the pre-coded vectors into a number of  $LM \times L$  subvectors; and

creating an  $M \times M$  diagonal matrix  $D_{s_g, k} = \text{diag}\{\Theta_{M \times (k-1)+1}^T \mathbf{s}_g, \dots, \Theta_{M \times k}^T \mathbf{s}_g\}$ , where  $k=1 \dots L$  from the subvectors.

32. (Previously Presented) A method according to claim 31, further comprising:

interleaving the  $L$  submatrices from the  $G$  groups to generate an  $M \times Nc$  space-frequency matrix.

33. (Previously Presented) A method according to claim 32, wherein the space-frequency matrix provides  $MNL$  channel diversity, while preserving a code rate of 1 for any number of the transmit antennae  $M$ , receive antennae  $N$  and channel tap(s)  $L$ .

34. (Previously Presented) A method according to claim 30, wherein the space-frequency matrix provides  $MNL$  channel diversity, while preserving a code rate of 1 for any number of the transmit antennae  $M$ , receive antennae  $N$  and channel tap(s)  $L$ .

35. (Previously Presented) An apparatus comprising:

a diversity agent:

to receive content for transmission from a wireless communication system having  $M$  transmit antennae and  $N$  receive antennae and  $Nc$  subcarriers, where  $Nc \gg M, N$ , the received content for transmission via a multicarrier wireless communication channel of the wireless communication system, wherein the received content is a vector of input symbols ( $\mathbf{s}$ ) of size  $Nc \times 1$ , and wherein the  $Nc$  subcarriers is the number of subcarriers of the multicarrier wireless communication channel;

and

to generate a rate-one, space-frequency code matrix from the received content for transmission on the multicarrier wireless communication channel from more than two of the  $M$  transmit antennae by dividing the vector of input symbols into a

number  $G$  of groups to generate subgroups and multiplying at least a subset of the subgroups by a constellation rotation precoder to produce a number  $G$  of pre-coded vectors ( $\mathbf{v}_g$ ), wherein successive symbols from the same group transmitted from the same antenna are at a frequency distance that is multiples of  $MG$  subcarrier spacings.

36. (Previously Presented) An apparatus according to claim 35, the diversity agent further comprising:

a space-frequency encoding element, responsive to the pre-coder element, to divide each of the pre-coded vectors into a number of  $LM \times I$  subvectors, and to create an  $M \times M$  diagonal matrix  $D_{s_g, k} = \text{diag}\{\Theta_{M \times (k-1)+1}^T \mathbf{s}_g, \dots, \Theta_{M \times k}^T \mathbf{s}_g\}$ , where  $k=1 \dots L$  from the subvectors.

37. (Previously Presented) An apparatus according to claim 36, wherein the space-frequency encoding element interleaves the  $L$  submatrices from the  $G$  groups to generate an  $M \times Nc$  space-frequency matrix.

38. (Previously Presented) An apparatus according to claim 37, wherein the space-frequency matrix provides  $MNL$  channel diversity, while preserving a code rate of 1 for any number of the transmit antennae  $M$ , receive antennae  $N$  and channel tap(s)  $L$ .

39. (Previously Presented) An apparatus according to claim 35, wherein the space-frequency matrix provides  $MNL$  channel diversity, while preserving a code rate of 1 for any number of the transmit antennae  $M$ , receive antennae  $N$  and channel tap(s)  $L$ .

40. (Previously Presented) A wireless communication system comprising:

a number  $M$  of omnidirectional antennas, wherein  $M$  comprises more than two omnidirectional antennas;

a number  $N$  of receive antennae;

a number  $N_c$  of subcarriers of a multicarrier wireless communication channel of the wireless communication system, where  $N_c \gg M, N$ ; and

a diversity agent:

to receive content for transmission via the multicarrier wireless communication channel, wherein the received content is a vector of input symbols ( $\mathbf{s}$ ) of size  $N_c \times 1$ , and to generate a rate-one, space-frequency code matrix from the received content for transmission on the multicarrier wireless communication channel from at least a subset of the  $M$  omnidirectional antennas by dividing the vector of input symbols into a number  $G$  of groups to generate subgroups and multiplying at least a subset of the subgroups by a constellation rotation precoder to produce a number  $G$  of pre-coded vectors ( $\mathbf{v}_g$ ), wherein successive symbols from the same group transmitted from the same antenna are at a frequency distance that is multiples of  $MG$  subcarrier spacings.

41. (Previously Presented) A wireless communication system according to claim 40, the diversity agent further comprising:

a space-frequency encoding element, responsive to the pre-coder element, to divide each of the pre-coded vectors into a number of  $LM \times 1$  subvectors, and to create an  $M \times M$  diagonal matrix  $D_{s_g, k} = \text{diag}\{\Theta_{M \times (k-1)+1}^T \mathbf{s}_g, \dots, \Theta_{M \times k}^T \mathbf{s}_g\}$ , where  $k=1 \dots L$  from the subvectors.

42. (Previously Presented) A wireless communication system according to claim 41, wherein the space-frequency encoding element interleaves the  $L$  submatrices from the  $G$  groups to generate an  $M \times N_c$  space-frequency matrix.

43. (Previously Presented) A wireless communication system according to claim 42, wherein the space-frequency matrix provides  $MNL$  channel diversity, while preserving a code rate of

1 for any number of the omnidirectional antennas  $M$ , receive antennae  $N$  and channel tap(s)  $L$ .

44. (Previously Presented) A wireless communication system according to claim 40, wherein the space-frequency matrix provides  $MNL$  channel diversity, while preserving a code rate of 1 for any number of the omnidirectional antennas  $M$ , receive antennae  $N$  and channel tap(s)  $L$ .

45. (New) The method of claim 30, wherein the diversity agent comprises an encoder to generate the rate one, space-frequency code matrix from the received content using a space frequency code, wherein the method further comprises:  
the encoder applying the space frequency code to the received content to generate the rate-one, space-frequency code matrix, wherein the space frequency code only consumes one multicarrier communication channel block duration.

46. (New) The method of claim 45, further comprising:  
transmitting the space frequency code in one OFDM block duration.

47. (New) The method of claim 30, further comprising:  
passing the rate-one, space-frequency code matrix as encoded content from the diversity agent to one or more inverse discrete Fourier transform (IDFT) elements; and  
transforming the encoded content from frequency domain into time domain content.

48. (New) The method of claim 47, wherein a quantity of the IDFT elements is commensurate with a quantity of the  $M$  transmit antennae.

49. (New) The method of claim 48, further comprising:  
passing the time domain content to one or more cyclical prefix insertion (CPI) elements to introduce a cyclical prefix or a guard interval prior to transmission via the  $M$  transmit

antennae.